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1 Wireless power transfer using relay resonators

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This paper presents an advanced design configuration of a wireless power transfer system using 6 7 overlapping relay coil techniques for free loading position. The work undertaken investigates the tuning position of prototype relay coils in a horizontal configuration in order to evaluate power 8 9 transfer efficiency versus increasing operating distance between the transmitter and receiver coils. A prototype relay coil system was evaluated to determine the optimum distance between the 10 transmitter and receiver. Finite element magnetic simulation was used to appraise the magnetic 11 field distribution and power transfer efficiency with respect to a strongly coupled magnetic 12 resonance condition. Experimental and simulation results analysis suggest that the proposed design 13 could achieve 85%–90% efficiency within a 4 cm operational distance. Finally, the experimental 14 results were analyzed and compared with the simulation results. *Published by AIP Publishing*. 15 https://doi.org/10.1063/1.5022032

Wireless Power Transfer (WPT) is now ubiquitous fol-16 lowing on from developments at MIT (Massachusetts 17 18 Institute of Technology), i.e., approximate 60 W of power was transferred between two 60 cm diameter coils which 19 were separated by a distance of 2 m.¹ WPT systems are now 20 used for powering electronic devices such as wireless charg-21 22 ing, radio frequency identification devices (RFIDs), sensor networks, medical implanted devices (cardiac pacemakers), 23 electric vehicles, maglev trains, and many more.²⁻⁴ 24

WPT can be classified as non-radiative or radiative 25 26 based on the system implemented in the power transfer process. The non-radiative type refers to near-field power trans-27 fer and it is applied in short or medium ranged applications 28 where the operational distance is below the transmitted 29 30 signal wavelength. Additionally, the transmitter-receiver dis-31 tance of the short-ranged application is less or equal to the diameter of the transmitter coil. Capacitive and inductive 32 coupling power transfer methods are forms of the short-33 ranged applications. The inductive coupling technique 34 35 function on the principle of magnetic induction, i.e., a transmitting coil induces an EMF within the receiver coil via a 36 magnetic field variation. The operational frequencies of a 37 WPT system depend on the inductance and capacitance val-38 ues of the coils. For capacitive coupling, energy is trans-39 ferred through an electric field, the quantity of energy 40 transferred being directly proportional to the frequency.⁵ 41 For medium range applications, the transmitter-receiver dis-42 tance varies between one to ten times the diameter of the 43 44 transmitter coil with an operational frequency range of 10 kHz-200 MHz. An inductively coupled system using mul-45 tiple coils as advantages over other systems (non-multiple 46 coils) by the extension of the power transfer distance 47 between the transmitter to receiver. The coupled magnetic 48 49 resonance system (CMRS) and inductive power transfer method are forms of the medium ranged applications.^{5,6} The 50 MIT research group described the detailed working principle 51

of CMRS system.⁷ This radiative type is known as far-field 52 power transfer. However, the operational distance is two 53 times higher than the transmitted signal wavelength. Its 54 implementation is based on electromagnetic wave propaga-55 tion in far field distance usually in the range of kilometers. 56 Essentially, it exists in two forms: directive and non-57 directive. Currently, the directive form is used for remotely 58 powering electric vehicles (evs), and the non-directive form 59 is used in applications such as power transfer between omni-60 directional RF broadcast and portable devices, transfer of 61 optical power through laser beams, and other applications.⁶ 62 There have been several studies over the last few years to 63 enhance the efficiency of the WPT system for short, medium, 64 and long distances using relay resonators between the trans-65 mitter and receiver coils.⁵⁻²⁴ For example, Choi and Lee⁵ 66 investigated the optimal position of relay coils from the 67 tested measured results without verifying the experimental 68 results with theory or simulation. In another paper, Ahn and 69 Hong³ demonstrated the WPT for multiple loads over vari-70 ous distances and verified the optimum conditions with mea-71 sured results. Park et al.⁸ discussed the magnetic field 72 repeater concept for the maximum power transfer for the 73 wireless system and provided the guidelines to select the 74 optimum repeater positions and number between the trans-75 mitter and receiver. They verified their concept experimen-76 tally and found that this concept could double the coupling 77 coefficient between the transmitter and receiver circuit. 78 However, it was observed that the coupling was not consid-79 ered for nonadjacent and adjacent cases. Zhang et al.¹⁰ stud-80 ied relay coupling but did not demonstrate the optimal power 81 transfer efficiency at the resonance frequency due to the 82 effect of nonadjacent coupling. Therefore, prior research in 83 this field discusses the theoretical background and investi-84 gates the performance of a test prototype without magnetic 85 simulation or verification of experimental results with simu-86 lation. An exception to this is the work of Zhang et al.¹¹ who 87 simulated the magnetic WPT system using simple transmit-88 ter and receiver coils and verified the simulated results with 89 measured results. However, Zhang *et al.*¹¹ did not use relay 90

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91 resonators or repeaters to extend the wireless power transfer distance between the transmitter and receiver. Magnetic simu-92 lation for a magnetic WPT system is necessary to be able to 93 appreciate the optimum size and flux linkage of the transmitter 94 95 and receiver resonators as the magnetic coupling and the resonance condition are the main factors in achieving the maxi-96 mum power transfer. This paper presents the design concept of 97 an overlapping relay resonator structure in horizontal configu-98 ration for WPT as opposed to the previous nonoverlapping 99 technique of vertical and horizontal relay resonators. This 100 design is particularly beneficial for low power medical applica-101 tions and also for free loading mobile charging applications. In 102 103 development, this design concept was simulated using the FEA magnetic simulation software package to aid understand-104 ing of the flux linkage and magnetic coupling factor of the 105 structure. An experimental circuit prototype was built and 106 107 tested using off-shelf components in order to acquire experimental data relating to the performance and operational wire-108 less power transfer distance limit. The simulated power 109 transfer efficiency with the variation of distance was compared 110 and verified. Finally, the efficiency comparison between this 111 112 study and other prior studies are discussed with conclusions

$$R_{TX} = R_{tx} + R_s \quad R_{RX} = R_{rx} + R_L$$

A typical induction coupling WPT consists of a trans-113 mitter and a receiver circuit, as shown in Fig. 1. The circuit 114 generates and transfers electrical energy between two reso-115 nant coils through varying magnetic fields. Since both coils 116 are closely coupled and operate at the same resonant fre-117 quency, high energy transfer can be achieved with small 118 leakage. This magnetic resonance coupling can be applied 119 between one transmitting coil and many receiving coils for 120 concurrent charging the multiple devices.⁷ The resonant fre-121 quency of the system can be calculated from $f_0 = \frac{1}{2\pi\sqrt{LC}}$, 122 where L and C represent the inductive and capacitive quanti-123 ties of the circuit. k is the coupling coefficient between the 124 transmitter circuit and receiver circuit. V_s is the source of 125 power for the transmitter circuit, R_s is the resistance of the 126 source, R_{tx} is the resistance in the transmitter circuit, R_{rx} is 127 the resistance in the receiver circuit, R_L is the load, C_1 is the 128 series connected transmitter capacitance, C_2 is the series 129 connected receiver capacitance, L_{TX} is the inductance in the 130 131 transmitter circuit, and L_{RX} is the inductance in the receiver circuit. The system's mutual inductance, i.e., where the mag-132 netic field generated by a coil causes voltage induction in the 133 coil adjacent, is found via¹² 134

$$M = k \sqrt{L_{TX} L_{RX}},\tag{1}$$

where the coupling coefficient, k, between the adjacent coils has been defined as¹⁸



FIG. 1. Equivalent circuit model of the transmitter-receiver WPT system.

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$$k = \frac{1}{\left|1 + 1.6\left(\frac{h^2}{r_l * r_r}\right)\right|^{\frac{3}{2}}},$$
 (2)

where h is the center to center distance between adjacent 137 coils, r_t is the transmitter coil radius, and r_r is the receiver 138 coil radius. It is necessary to match the load impedance with 139 the source impedance to transfer maximum power at the load 140 end. Since both the magnetic field strength and coupling fac- 141 tor gradually decrease with the distance, the power transfer 142 distance of a magnetically coupled WPT system will depend 143 on the size and properties of the magnetic material and the 144 coil dimensions. High Q-factor transmitter and receiver coils 145 could compensate the low magnetic coupling to increase the 146 power transfer distance to a certain degree. However, an 147 excessive Q-factor could represent excessive reactance for 148 each load resulting in an increase of the existing magnetic 149 field which could pose a risk to body tissue. The relay coils 150 implementation as shown in Fig. 2 could be a suitable option, 151 i.e., to improve the magnetic coupling over a longer distance 152 which will increase the power transfer distance from the 153 transmitter to the receiver end. 154

It can be seen from the schematic circuit in Fig. 2 that 155 the typical relay coil WPT comprises of basic transmitter 156 and receiver circuit with n number of resonators. 157

The transformer coupling principle proposes that the 158 product of capacitance and inductance in circuit must be the 159 same to have a resonant circuit¹⁰ 160

$$f_{TX} = f_1 = f_2 = f_3 = f_4 = f_5 = f_6 = f_7 = f_{RX}, \quad (3)$$

$$L_{TX} \times C_1 = L_A \times C_3 = L_A \times C_3 = \dots L_{RX} \times C_2.$$
(4)

The wireless system power transfer efficiency is described as 161 the ratio of the power dissipated at the load to the power 162 input to the transmitter and is derived by summing the total 163 dissipated power in the coils and the load.¹² 164

Applying Kirchhoff's voltage law to solve the circuit, 165 the power transfer efficiency is given by⁹ 166

$$\eta = \frac{P_L}{P_{in}},\tag{5}$$

$$=\frac{\frac{1}{2}R_{L}|I_{n}|^{2}}{\frac{1}{2}R_{tx}|I_{tx}|^{2}+\frac{1}{2}R_{a}|I_{1}|^{2}+\dots+\frac{1}{2}R_{n-1}|I_{n-1}|^{2}+\frac{1}{2}(R_{L}+R_{n})|I_{n}|^{2}},$$

$$=\frac{R_{L}}{R_{tx}\left|\frac{I_{tx}}{I_{n}}\right|^{2}+R_{a}\left|\frac{I_{1}}{I_{n}}\right|^{2}+\dots+R_{n-1}\left|\frac{I_{n-1}}{I_{n}}\right|^{2}+(R_{L}+R_{n})}.$$
(6)



FIG. 2. Equivalent circuit model of the WPT system with n resonators.

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FIG. 3. Practical setup of the horizontal design for wireless power transfer.

167 The current ratio $\frac{I_{tx}}{I_n} \dots \frac{I_{n-1}}{I_n}$ can be realized using inverse 168 matrix on the resolved matrix form of the circuit. Therefore, 169 $\eta = f(C_3, \dots, C_2, R_L)$. It should be noted that C_1 and R_s are 170 related to impedance matching and not the power transfer 171 efficiency. So, for maximum power transfer analysis, realiza-172 tion of C_3, \dots, C_2, R_L are critical.

Figure 3 shows the prototype of the relay coils WPT system which was built using off-shelf components and characterized experimentally. The design is implemented using seven overlapping relay coils alternatingly placed at approximately center to center distances as shown.

Essentially, 3-D FEA magnetic transient simulation
using Infolytica package was carried out to determine the
flux linkage and power transfer efficiency over the distance.
Finally, the measured results were analyzed and discussed
with the simulation results.

The components used to implement the WPT system 183 include commercial Qi transmitter and receiver. The trans-184 185 mitter circuit consists of an inverter circuit integrated with a ferromagnetic transmitter coil. The maximum input voltage 186 of the transmitter circuit is 5 V (dc) recommended by the 187 manufacturer, and the output of the inverter will provide 188 50 mV (rms) sinusoidal voltage to the transmitter coil for 189 magnetic induction.¹³ The receiver comprises of a copper 190 coil, a ferromagnetic layer, and the rectifier circuit;^{13–15} relay 191 coils are made of copper having 20 number of turns which 192 have been tuned at a resonance frequency to maximize the 193 power transfer. The operation of wireless power transfer sys-194 tem at the resonant frequency is necessary to achieve the 195 maximum output power and efficiency. The resonant fre-196 quency of the transmitter and receiver circuit is 220 kHz, and 197 the relay coils are tuned with this frequency to achieve the 198 199 maximum efficiency. The distance of the each relay coil 200 from the transmitter circuit, as shown in Fig. 3, is the



FIG. 4. 3-D transient simulation model of the WPT system.

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TABLE I. Coil parameters used in simulation.

	No. of turns	Resistance $(m\Omega)$	Inductance (µH)	Inner radius (cm)	Outer radius (cm)
Transmitter coil	10	180	24	1.2	2.3
Relay coils	20	232	24	1.5	3
Receiver coil	35	135	8	1.1	2.2



FIG. 5. Effective flux linkage for all the coils.



FIG. 6. Flux linkage and calculated mutual inductance over operational distance of the relay resonators.



FIG. 7. Relay coil inductance values and quality factor with the variation of frequency.

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FIG. 8. Measured and simulated output power and efficiency of the relay coils.

optimum distance for this configuration, i.e., that would 201 allow connectivity and transmission to subsequent coils. 202

Figure 4 presents the descriptive model of WPT system 203 with relay coils which has been used in the finite element 204 magnetic simulation. The first coil integrated with the ferrite 205 material is the transmitter coil, and the subsequent coils are 206 the relay coils. Table I shows the coil parameters used in the 207 simulation. The transmitter coil is driven by a sinusoidal 208 209 voltage source with the magnitude of 50 mV (rms) at 50 Hz frequency. Figure 5 shows the sinusoidal flux variation over 210 the time period of the WPT system for individual coils. In 211 order to understand the flux linkage over the operational dis-212 tance, the rms values of the flux linkages were plotted, as 213 shown in Fig. 6. The distance between the center of the 214 transmitter and receiver coils varied due to the change in 215 position of the receiver coil across the relay coils resonator. 216 The distances represent the points where the different relay 217 coils are placed. It can be seen from the graph that the flux 218 linkage decreases near exponentially over distance with the 219 flux linkage dropping by 65% at the 4th coil from the trans-220 mitter, at the 7th coil, and beyond power transfer is not 221 achievable as the flux linkage reduction is 92%. Initially, the 222 inductance and the quality factor (Q) of the relay coils were 223 measured with the variation of frequencies. Figure 7 shows 224 225 the measured inductance and the calculated Q factor for 20 turn coils. The quality factor, Q, is determined using the 226



FIG. 9. Measured and simulated output power.

formula O = $\frac{\omega L}{P}$ from the measured inductance and resistance 227 values. 228

An LCR meter was used to measure the inductance and 229 resistance of the coils at different frequencies. The coil 230 exhibits a maximum quality factor 150 at 220 kHz frequency. 231 Several relay coils with different number of turns were used 232 in the experiment to evaluate the quality factor variation 233 with frequency. 234

The experimental results suggest that the coil perfor- 235 mance is critically affected by the number of turns and the 236 wire gap of the coil. The coupling factor (k) and the mutual 237 inductance of the WPT system have been evaluated using 238 Eqs. (2) and (1), respectively, to evaluate how they affect the 239 performance of power and efficiency over the distance across 240 the relay coils between 2 cm and 12 cm. It can be seen from 241 Fig. 6 that the mutual inductance is significantly reduced 242 over the operational distance. 243

The transmitter circuit is supplied by a 5 V DC source, 244 and the input/output voltage and power of the WPT system 245 have been measured to characterize the prototype. The mea- 246 sured output power for 2.5 cm and 12 cm distance are 2.5 W 247 and 0.9 W, respectively. Figure 9 shows the measured and 248 simulated normalized output power and power transfer effi- 249 ciency against the different distances/position of the receiver 250 coil across the relay resonators. Measured and simulated out- 251 put powers were normalized based on the maximum simu- 252 lated output power. 253

TABLE II. Comparison with previous works.

Reference	Coil dimension: r_t , and r_r (mm)	Operating frequency (MHz)	Efficiency without relay coils (%)	Efficiency with relay coils (%)	Operational distance (mm)
Rakhyani et al. ²⁵	32, 11	0.70	28	77%, 2 relay coils	30
Rashid ²⁶	130, 130	38.15	4.5	25%, 1 relay coil	1000
				75%, 2 relay coils	
Zhang <i>et al.</i> ¹⁰	81, 81	8	10	46%, 2 relay coils	300
Bhutada et al. ²⁷		4-12	40	93%, 2 relay coils	500
Zhu et al. ²⁸	23, 22	0.25	15.9	36.3%, 2 relay coils	20
Wang et al. ²⁰	8.5, 6	200		30%, 2 relay coils	7
This work	23, 22	0.22	35	85%–90%, 7 overlapping relay coils	27

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Figure 8 shows the comparison between measured and simulated power transfer efficiency for the relay coil system. It can be seen that the simulated results are closely matched with the measured results and that the efficiency drops from 90% to 35% over a 10 cm distance. The receiver output power was also measured without a relay coil to evaluate the operational distance and efficiency with and without a relay coil.

Figure 9 clearly indicates that without a relay coil the efficiency will drop from 85% to 0% within very short operational distance of between 2 cm and 3 cm.

Both mutual inductance and the power transferred both decrease with increasing the distance between the transmitter and receiver coils. It can be seen that high power transfer can be achieved with a low source impedance which will provide the low I^2R_s loss in the system.

This study has investigated the tuning position of prototype overlapping relay coils in a horizontal design for distance extension of WPT and evaluated power transfer efficiency with the increase in operating distance between the transmitter and receiver coils.

274 Table II summarizes the comparison between this study and previous research²⁰⁻²³ with and without relay coils and 275 demonstrated that the relay coil WPT system will increase the 276 efficiency significantly compared to a system without a relay 277 278 coil. However, the efficiency and the operational distance 279 depend on the coil dimension and magnetic material of the transmitter and receiver coils. It can be seen from Table II that 280 the overlapping relay resonators show the highest WPT effi-281 ciency compared to the previous findings. Additionally, the 282 283 overlapping relay coils design of the WPT system would 284 enhance a wide range of applications such as implantable medical devices, i.e., cardiac pacemaker, nerve simulator, and reti-285 nal prosthesis with other applications such as wireless charging 286 and wearable sensing devices.¹⁶ 287

The design configuration is simulated using the FEA 288 magnetic software, and the simulated results have been 289 analyzed and compared with the measured results. The 290 measured results suggest that the prototype is capable of 291 achieving 85%–90% efficiency within a 4 cm distance 292 from the transmitter coil. The relay coil resonators have 293 294 been built and tested for the purpose of extending power transfer. This is particularly beneficial in health and medi-295 cal applications and also in mobile charging applications 296 due to small power requirements. These areas are undergo-297 ing further study. The eventual goal is to optimize the relay 298 299 coils and transmitter/receiver coils using the magnetic simulation software with the horizontal array configuration. 300 Furthermore, this horizontal configuration is compared 301 with the vertical array configuration to optimize the 302 303 design.

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